

(12) INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(19) World Intellectual Property Organization  
International Bureau



(43) International Publication Date  
2 August 2001 (02.08.2001)

PCT

(10) International Publication Number  
**WO 01/56146 A1**

(51) International Patent Classification<sup>7</sup>:

H03F 1/32

(81) Designated States (*national*): AE, AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, CA, CH, CN, CR, CU, CZ, DE, DK, DM, EE, ES, FI, GB, GD, GE, GH, GM, HR, HU, ID, IL, IN, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MA, MD, MG, MK, MN, MW, MX, NO, NZ, PL, PT, RO, RU, SD, SE, SG, SI, SK, SL, TJ, TM, TR, TT, TZ, UA, UG, US, UZ, VN, YU, ZA, ZW.

(21) International Application Number:

PCT/EP00/00579

(22) International Filing Date: 26 January 2000 (26.01.2000)

(25) Filing Language:

English

(26) Publication Language:

English

(71) Applicant (*for all designated States except US*): NOKIA NETWORKS OY [FI/FI]; P.O. Box 300, FIN-00045 Nokia Group (FI).

(72) Inventor; and

(75) Inventor/Applicant (*for US only*): MASHHOUR, Ashkan [FR/GB]; Stanhope Road, Yorktown Industrial Estate, Camberley, Surrey GU15 3BW (GB).

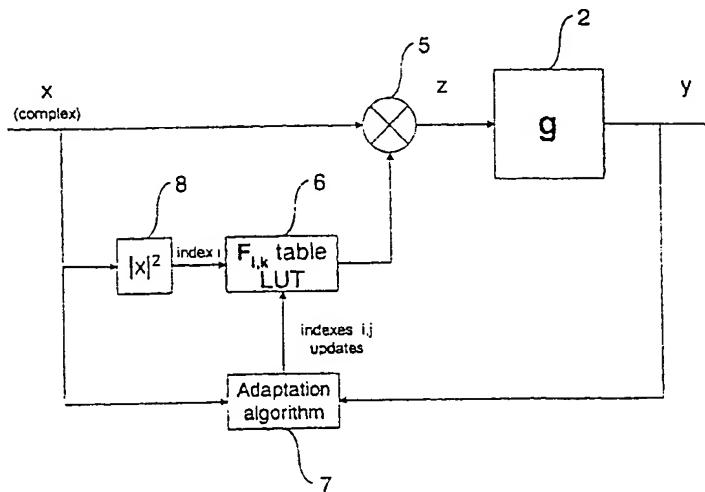
(84) Designated States (*regional*): ARIPO patent (GH, GM, KE, LS, MW, SD, SL, SZ, TZ, UG, ZW), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE), OAPI patent (BI, BJ, CF, CG, CI, CM, GA, GN, GW, ML, MR, NE, SN, TD, TG).

Published:  
— with international search report

*For two-letter codes and other abbreviations, refer to the "Guidance Notes on Codes and Abbreviations" appearing at the beginning of each regular issue of the PCT Gazette.*

(74) Agents: PELLMANN, Hans-Bernd et al.; Tiedtke-Bühling-Kinne et al., Bavariaring 4, D-80336 München (DE).

(54) Title: METHOD AND SYSTEM FOR COMPENSATING NON-LINEARITIES AND TIME-VARYING CHANGES OF A TRANSFER FUNCTION ACTING ON AN INPUT SIGNAL



**WO 01/56146 A1**

(57) Abstract: The described method and system are adapted to reduce the error between an ideally expected output signal and the actual output signal. The proposed adaptation algorithm is able to minimise, for instance in a system with a given transfer function, the error  $y-x$  between  $y=g(f(x))$  and  $x$ , where  $g$  is an unknown and/or time-varying function and  $f$  the adaptive function for which the characteristic is changed to track  $g$ . The proposed adaptation algorithm updates not only the transfer function  $f$  at the current input value  $x$ , but also the transfer function  $f$  at other points corresponding to different input values. One of the applications for such an algorithm is digital predistortion where a transmitter's non-linear characteristic needs to be linearised, in an adaptive manner, since the characteristic exhibits slow changes with temperature, bias, ageing or the like.

Method and system for compensating non-linearities and time-varying changes of a transfer function acting on an input  
5 signal

Description

10 Field of the Invention

The invention is directed to a method for compensating non-linearities and time-varying changes of a transfer function acting on an input signal. This input signal is first subjected  
15 to a first, adaptive transfer function and thereafter to a second, unknown and varying transfer function to generate the actual output signal. When the second transfer function varies, the first transfer function is updated for compensating these changes. The described method and system are therefore adapted  
20 to reduce the errors between an ideally expected output signal and the actually generated output signal.

Background of the Invention

25

For example, in the field of mobile communications, there is sometimes the need to linearise a transmitter's non-linear characteristic, in an adaptive manner, when the characteristic exhibits slow changes caused by temperature, bias, ageing or  
30 the like. One of the existing compensating possibilities is the digital predistortion of the input signal before applying same to the transmitter's power amplifier. Thereby, the error between an ideally expected output signal and the output signal actually generated in response to a current input signal can be  
35 minimised. This need for compensation not only occurs in the field of mobile communications but also in other fields where a

transfer function varying in an unknown manner is to be compensated by adapting an adaptive transfer function.

When the characteristic is changing, it takes some time until  
5 the system has been adapted to the new situation. It is desirable to perform this adaption as swiftly as possible. However, normally, a great number of iterations is necessary until the system has actually been adapted to the new characteristic.

10

#### Summary of the Invention

It is therefore an object of the present invention to provide a  
15 method and system which are able to swiftly adapt to a changing transfer function and thereby quickly minimise an error between the actual output signal caused by a current input signal, and an ideally expected output signal.

20 The invention provides a method for compensating deviations of an unknown transfer function from an expected transfer function, and/or for compensating time-varying changes of a transfer function acting on an input signal, for minimising errors of an output signal generated in dependence on the input  
25 signal, which input signal is subjected to a first, adaptive transfer function and to a second, varying transfer function to generate the output signal, the first transfer function being updated for compensating deviations of the second transfer function from an expected transfer function and/or for  
30 compensating changes of the second transfer function, wherein, when updating one point of the first transfer function for a current input signal value, the first transfer function is also updated for at least one other point corresponding to a different input signal value.

35

The invention furthermore provides a system for compensating an unknown and/or varying transfer function acting on an input

- signal, for minimising errors of an output signal generated in dependence on the input signal, which input signal is applied via a first, adaptive compensating means having a first, adaptive transfer function, to a second means having a second, 5 unknown and/or varying transfer function to generate the actual output signal, the first transfer function being updated for correcting deviations of the second transfer function from an expected (ideal, wanted) transfer function by means of a processing means, wherein the processing means is adapted to 10 update, when updating one point of the first transfer function for a current input signal value, the first transfer function also at least one other point corresponding to a different input signal value.
- 15 Due to this multi-point-correction of the first transfer function, the described method and system are able to quickly adapt the first transfer function to changes of the second transfer function and/or to quickly adapt the overall transfer function to the desired one. Thus, the apparatus swiftly 20 converges to the new (or desired) condition, with a drastic reduction of the necessary adaption time. The occurrence of errors between an output signal ideally to be expected for a given input signal, and the actually generated output signal is therefore limited to a very short time interval after a 25 variation (or the first use) of the second transfer function. Furthermore, the error deviations within this time interval are reduced to smaller values.
- 30 Brief Description of the Drawings

The present invention will be more readily understood upon referring to the following description of preferred embodiments when read in conjunction with the accompanying drawings. In the 35 drawings,

Fig. 1 shows a basic configuration of a functional system,

Fig. 2 shows a chart for explaining the algorithm used in the present invention,

5 Fig. 3 shows the system of Fig. 1 in a practical implementation,

Fig. 4 shows an iterative updating of two values of a transfer function,

10 Fig. 5 illustrates the updating of a plurality of values of a transfer function,

Fig. 6 shows an embodiment of the invention implemented as an adaptive predistorter and amplifier,

15

Fig. 7 shows an example of the magnitude and phase response of function  $g$ ,

20 Fig. 8 illustrates an example of the ideally expected and actually generated output signals for a non-linearised function  $g$ ,

Figs. 9 and 10 show simulation results for the presented algorithm and known algorithms,

25

Fig. 11 illustrates F values calculated at the end of the simulation,

30 Figs. 12 and 13 show the in-phase and quadrature-phase components of input and output signals,

Figs. 14 and 15 illustrate the different convergence behaviour of the presented and known methods for different variations of the transfer function  $g$ , and

35

Figs. 16 and 17 show the results of three different updating procedures using the linear and LMS methods, resp.

Detailed Description of Preferred Embodiments of the Invention

5 Generally, the described adaptation algorithm is a method of minimising the error  $|y-x|$  between  $y=g(f(x))$  and  $x$ , where  $g$  is an unknown function and  $f$  the adaptive function for which the characteristic is changing to track  $g$ . Ideally,  
10  $f=g^{-1}$ . Unlike current algorithms, it updates not only the transfer function  $f$  at the current input value  $x$ , but also the transfer function  $f$  at other points (corresponding to different input values  $x_{update\dots}$ ). One of the applications for such an algorithm is digital predistortion where a transmitter's non-linear characteristic needs to be linearised, in an adaptive manner since the characteristic exhibits slow changes with  
15 temperature, bias, ageing....

First, the new algorithm used in the method and system will be mathematically described.

20

Let us consider an unknown function  $g$  defined as:

$$g: D_g \subseteq C \rightarrow C$$

$$x \rightarrow g(x)$$

25  $C$  designates the complex domain, and can be extended to an  $n$ -dimension domain;  $D_g$ : definition domain of  $g$ .

The adaptive function  $f$  is defined as:

$$\begin{array}{ll} f: D_f \subseteq C \rightarrow C & D_f : \text{definition domain of } f \\ 30 & x \rightarrow f(x) \quad C: \text{complex numbers} \end{array}$$

35 The functions (transfer functions)  $f$  and  $g$  are performed in the order shown in Fig. 1. Fig. 1 illustrates a functional system having a compensating means 1 to which an input signal  $x$  is supplied at a first input terminal. The compensating means 1 generates an output signal  $z$  in accordance with its transfer

function  $f$  (correspondence between  $x$  and  $z$ ) and the value of  $x$ . The signal  $z$  is input into a second means 2 which outputs the output signal  $y$  and has a transfer function  $g$  which may change over time. The output signal  $y$  is fed back to a second input 5 terminal of the compensating means 1. As the overall transfer function is to be maintained constant, any change of the transfer function  $g$  or deviation from the ideal transfer function will be compensated by appropriately adapting the transfer function  $f$  so as to maintain the desired relationship 10 between  $x$  and  $y$ . The means 1 may for example be an adaptive predistorter whereas means 2 may be a power amplifier. When the desired relationship between  $x$  and  $y$  is equity (amplification gain factor = 1) for example, any deviation of the gain of the amplifier 2 from the value 1 will be compensated by setting the 15 amplification characteristic of means 1 to the reciprocal value.

The algorithm attempts, in the present example with unity gain (gain factor = 1), to minimise the error  $|y-x|$  between the 20 output  $y$  and the input  $x$ , in an iterative process (see Fig. 2).

Fig. 2 is a diagram showing the characteristic (transfer function)  $g$  and the desired linearised overall response characteristic 3 ( $x=y$ ). The horizontal axis represents the 25 absolute value of the output  $z$  of means 1 whereas the vertical axis represents the final output  $y$  (absolute value).

In the following,  $f_k(x)$  denotes  $f(x)$  at iteration  $k$ , where an iteration is the process of updating simultaneously one or more 30 values defining the function  $f$ . It will also be written as:

$$f_k(x) = x \cdot F_{k,x} \text{ (with } F_{k,x} \in C\text{)}.$$

At iteration 0 ( $k=0$ ), the transfer function  $f$  is set to 1:

$$\forall x \in C, F_{0,x} = 1 + j \cdot 0$$

35 (i.e.,  $\forall x \in C, f_0(x) = x$ ).

$$y = g(z) = g(f_k(x)) = g(x \cdot F_{k,x}) = x \cdot F_{k,x} \cdot G_{k,x} \quad (1)$$

with  $x \cdot F_{k,x} = f_k(x)$  and  $G_{k,x} = \frac{y}{z} = \frac{g(x \cdot F_{k,x})}{x \cdot F_{k,x}}$

Ideally, we would like:

5  $\forall x \in C, y = x$

substituting in (1), ..

$$\Leftrightarrow x \cdot F_{k,x} \cdot G_{k,x} = x \quad (2)$$

10  $\Leftrightarrow F_{opt} = \frac{1}{G_{k,x}}$  (defined as such) (3)

$F_{opt}$  would be the optimum value to use for  $F_{k,x}$  to satisfy (2). Unlike previously published algorithms (linear method, secant method, Least Mean Square (LMS) etc...), which have been used more specifically for predistortion systems), it is not theoretically correct to update the new value  $F_{k+1,x}$  with  $F_{opt}$  (even though adaptation factors are used to avoid instability). If  $F_{k+1,x}=F_{opt}$  would be given, (2) would become:  $x \cdot F_{k+1,x} \cdot G_{k,x} = x$  which is the desired result.

20 However, this does not take into account the fact that changing  $F_{k+1,x}$  would mean  $x \cdot F_{k+1,x} \neq x \cdot F_{k,x}$ , yielding a different input to  $g$  (see Fig. 2).

25 Let's define  $y_1$  and  $y_2$  as:

$$y_1 = y = g(f_k(x)) = g(x \cdot F_{k,x}) = x \cdot F_{k,x} \cdot G_{k,x}$$

$$y_2 = g(f_{k+1}(x)) = g(x \cdot F_{k+1,x}) = x \cdot F_{k+1,x} \cdot G_{k+1,x} \text{ (with } F_{k+1,x}=F_{opt} \text{ here)}$$

therefore  $y_2 \neq x$  because  $G_{k,x} \neq G_{k+1,x}$

30 What really ought to be updated is the value of  $f$  corresponding to a different input  $x_{update}$ , while  $F_{k+1,x}$  remains unchanged ( $F_{k+1,x}=F_{k,x}$ ).  $x_{update}$  is defined as:

$$f_{k+1}(x_{\text{update}}) = f_k(x)$$

$$\Leftrightarrow x_{\text{update}} \cdot F_{k+1,x_{\text{update}}} = x \cdot F_{k,x} \quad \text{with } F_{k+1,x_{\text{update}}} = F_{\text{opt}}$$

$$5 \quad \Leftrightarrow x_{\text{update}} = \frac{x \cdot F_{k,x}}{F_{\text{opt}}} \quad (4)$$

substituting for  $F_{\text{opt}}$  from (3) yields,

$$x_{\text{update}} = x \cdot F_{k,x} \cdot G_{k,x} \quad (5)$$

10

comparing (5) with (1),

$$\Leftrightarrow x_{\text{update}} = y \quad (6)$$

Therefore, now (iteration  $k+1$ ),  $x_{\text{update}}$  yields the same input to  
15  $g$  as was  $x$  at the previous iteration (iteration  $k$ ):

$$\begin{aligned} f_{k+1}(x_{\text{update}}) &= f_k(x) \\ \Updownarrow \\ g(f_{k+1}(x_{\text{update}})) &= g(f_k(x)) \end{aligned}$$

assuming  $x$  has not varied from iteration  $k$  to  $k+1$

$$\begin{aligned} \Leftrightarrow G_{k+1,x_{\text{update}}} &= G_{k,x} \\ \Leftrightarrow y_{\text{update}} &= g(f_{k+1}(x_{\text{update}})) = g(x_{\text{update}} \cdot F_{k+1,x_{\text{update}}}) = g(x_{\text{update}} \cdot F_{\text{opt}}) \\ 20 \quad \Leftrightarrow y_{\text{update}} &= g(x \cdot F_{k,x}) \quad \text{from (4)} \\ \Leftrightarrow y_{\text{update}} &= y \end{aligned} \quad (7)$$

Recalling (6) and having (7), the result is

$$25 \quad y_{\text{update}} = x_{\text{update}} \quad (8)$$

which corresponds to a linearised system at point  $x_{\text{update}}$   
(remember the aim was to have  $y=x$ ).

30 In this form however, the algorithm may present the danger of  
not updating  $f$  completely since the current  $f_{k,x}$  is pointing at

another value  $f_{k+1,x_{\text{update}}}$  to be updated. And once the latter has been updated, it will not point to another  $f_{k+1,x_3}$  ( $x_3 \neq x_{\text{update}}$ , determined in the same manner as  $x_{\text{update}}$  was obtained from  $x$ ) but merely to itself.

5

Bearing this in mind, the problem is preferably overcome by updating the current value  $f_{k+1,x}$  for the current input  $x$  too. The following overall solution is suggested:

10       $F_{k+1,x_{\text{update}}} = \mu \cdot F_{\text{opt}} \text{ or similar} \quad (\text{a})$

$F_{k+1,x} = \text{chosen algorithm (linear, LMS, secant etc...)} \quad (\text{b})$

where  $\mu \in \mathbb{R}$  (real numbers) and is constant ("adaptation factor").

15

It is important to note that the key point here is to update not only the  $F_{k,x}$  value corresponding to the current input  $x$  (i.e.  $F_{k+1,x}$  - this is done in (b)), but also  $F_{k,x_{\text{update}}}$  corresponding to a different value  $x_{\text{update}}$  defined by (4) (i.e.

20       $F_{k+1,x_{\text{update}}}$  - this is done in (a)). Hence, a different algorithm could be used to update  $F_{k,x}$  ( $F_{k+1,x}$ ), like LMS, the linear method, Recursive Least Squares (RLS) etc... In other words, the important part is (a), whereas any algorithm can be used for (b); the adjunction of (a) to (b) improves (b).

25

One could choose for (b):  $F_{k+1,x} = \frac{\alpha \cdot F_{k,x} + \beta \cdot F_{\text{opt}}}{2}$  with  $(\alpha, \beta) \in \mathbb{R}^2$ .

Note that, for example, if  $\alpha = \beta = \mu = 1$  and assuming the algorithm converges, the following holds true:

30       $F_{k,x} \xrightarrow{k \rightarrow \infty} F_{\text{opt}}$

$\mu, \alpha, \beta$  can be chosen as a function of the input  $x$ , and therefore be different for different  $x$  inputs.

In the foregoing explanation, the equation  $x=y$  was assumed as desired relationship between  $x$  and  $y$ . In practise, however, the present algorithm can also be applied, with appropriate changes, to a more complicated dependency of  $y$  on  $x$ , for 5 instance:  $y=\lambda \cdot x$  ( $\lambda \in \mathbb{C}$ , the previous description used  $\lambda=1$ ).

It can furthermore be envisaged to update adjacent indices to  $F_{k+1,x_{\text{update}}}$  as well (corresponding to other inputs), with appropriate weighting. This is because  $g$  can be assumed to have 10 a continuous derivative ( $g$  is „smooth“); doing this results in an even faster convergence (a particular illustration is discussed in the next section below).

One possible variation consists in updating  $f$  for all values 15 between  $x$  and  $x_{\text{update}}$ , by interpolation. A possible way of doing this is to use linear interpolation. Assuming  $x \leq x_{\text{update}}$ ,

$$F_{k+1,x_{\text{interpolate}}} = F_{k+1,x} + \frac{F_{k+1,x_{\text{update}}} - F_{k+1,x}}{x_{\text{update}} - x} \cdot (x_{\text{interpolate}} - x)$$

with  $x_{\text{interpolate}} \in [x, x_{\text{update}}]$

This can increase convergence speed quite considerably.

20 In the following, a practical implementation will be described where  $D_f$  is finite.

In practise, the above described algorithm may be implemented 25 in a computer program or in a DSP (digital signal processor), requiring digitisation. It is then suitable to limit the size of  $D_f$  and define  $f$  as a list of complex values in a table which are constantly updated. The table can be multi-dimensional and the addressing customised to the application (e.g.: one could 30 use the magnitude of the input value to address the table, or the square magnitude/power etc...).

The following example is an illustration of this particular case with a one-dimensional table of size  $m$ , addressed by the

magnitude of the input  $x$ . Since the table is finite, each input value  $x$  is associated with an index to the table (quantisation), fetching the corresponding value  $F_{i,k}$  in the table. For example:  $f_k(x) = x \cdot F_{i,k}$ ;  $f_k(x)$  can also be produced by interpolation.

Figure 3 shows a practical implementation of this system similar to same of Fig. 1 wherein the compensating means 1 is implemented as a table memory (here: one-dimensional look-up table) 4 which is addressed by the actual value of the input  $x$ . The values memorised in the table memory 4 are used for multiplication with the value of  $x$  for generating  $z$ , and are iteratively updated.

Figure 4 illustrates the update process and shows, in the left part, the table contents (values  $F_{1,k}$  to  $F_{m,k}$ ) at iterative step  $k$ , whereas the right part of Fig. 4 illustrates the table contents (values  $F_{1,k+1}$  to  $F_{m,k+1}$ ) at the next iterative step  $k+1$ . The arrows between the left and right parts of Fig. 4 indicate the double updating not only of  $F_{i,k+1}$  but also of  $F_{j,k+1}$ , corresponding to  $x$  and  $x_{\text{update}}$ , resp.

Applying the interpolation discussed above results in the diagram shown in Fig. 5. This can increase convergence speed considerably, however, it is computationally more complex and resource consuming. Fig. 5 illustrates the structure, wherein more than two  $F$  values are updated based on the present value of  $x$ . Similar to Fig. 4, Fig. 5 shows, in the left part, the table contents (values  $F_{1,k}$  to  $F_{m,k}$ ) at iterative step  $k$ , whereas the right part of Fig. 5 illustrates the table contents (values  $F_{1,k+1}$  to  $F_{m,k+1}$ ) after the next iterative step  $k+1$ . Here, in addition to the updating not only of  $F_{i,k+1}$  but also of  $F_{j,k+1}$  all values of  $F$  assigned to values of  $x$  lying between  $x$  and  $x_{\text{update}}$  are updated ( $F_{i-1,k+1} \dots F_{j+1,k+1}$ ), for example by interpolation. The plurality of arrows between the left and right half parts of Fig. 5 indicate this multi-updating based on one actual  $x$  value.

The algorithm described above has to be adapted to this application, but basically remains very similar. We can choose for instance to address the F table by the magnitude of the  
5 input x (affects look-up table resolution).

It is assumed that the input x has a known limited magnitude range:  $|x| \in [|x|_{\min}; |x|_{\max}]$ . The F table size is set accordingly to cover all the input values of x within this range. The total  
10 range of magnitude of input x is divided into subranges of possibly, but not necessarily equal size (it can for instance be a function of the input signal's statistical distribution ...), the number of subranges corresponding to the number m of values of F contained in the table. Therefore, each actual  
15 value of x addresses that table value F which is attributed to that subrange to which the present x belongs, for instance  $F_i$ . Mathematically:

for  $|x| \in [a_i, a_{i+1}]$ ,  $x \xrightarrow{\text{associated to}} \text{index } i \text{ in the F table corresponding to } F_{i,k}$   
with  $a_i \in \mathbb{R}^+$ , and  $i \in [1 \rightarrow m]$   
20 Hence,  $f_k(x) = x \cdot F_{i,k}$  (interpolation may be used here to determine  $f_k(x)$ ).

The index i can be determined as:

$$i = \text{floor}\left(\frac{(m-1) \cdot |x|^2}{|x|_{\max}^2}\right) + 1 \quad (9)$$

25 Equation (9) makes sure that the indices i range from 1 for  $|x|=0$  to  $i=m$  for  $|x|=|x|_{\max}$ .

Repeating the algorithm as it has been described in part 1:

$$y = g(f_k(x)) = g(x \cdot F_{i,k}) = x \cdot F_{i,k} \cdot G_{i,k,x}$$

30 with  $x \cdot F_{i,k} = f_k(x)$  and  $G_{i,k,x} = \frac{y}{z} = \frac{g(x \cdot F_{i,k})}{x \cdot F_{i,k}}$

$$F_{\text{opt}} = \frac{1}{G_{i,k,x}}$$

As this practical implementation of the system has a finite  $D_f$ , the additional index "i" is added which refers to the value  $F$  in the look-up table corresponding to cell "i" (see Fig. 4). "k" is the iteration number, whereas "x" is used as a referring index for the gain  $g$  at point  $x \cdot F_{i,k}$  which is the input to the transfer function  $g$ .

So  $x_{\text{update}} = \frac{x \cdot F_{i,k}}{F_{\text{opt}}}$  and the associated index in the F table for  $x_{\text{update}}$  is  $j$ , with  $j \neq i$ .

10

It must be ensured that  $|x_{\text{update}}|$  is within the range covered by the F table (i.e.,  $|x|_{\min} \leq |x_{\text{update}}| \leq |x|_{\max} \Leftrightarrow j \in [1 \rightarrow m]$ ).

15 If  $|x_{\text{update}}| \leq |x|_{\min}$  then  $j$  is preferably set to  $j=1$  and if  $|x|_{\max} \leq |x_{\text{update}}|$  then  $j$  is preferably set to  $j=m$ . Alternatively, the F table could simply not be updated at index  $j$  for  $j \notin [1 \rightarrow m]$ .

20 Finally, the following settings are selected, similar to the above discussed case:

$$F_{j,k+1} = \mu \cdot F_{\text{opt}} \quad (i)$$

$$F_{i,k+1} = \text{chosen algorithm (linear, LMS, secant etc...)} \quad (ii)$$

25

where  $\mu \in \mathbb{R}$  and is constant.

If  $i=j$ , one may choose either (i) or (ii) for the update.

30 This algorithm has been simulated in a system commonly referred to in the literature as predistortion.

Fig. 6 shows a predistortion block diagram. The input  $x$  is complex and has been represented as:

$$x(t) = \sin(\omega.t) + j \cdot \sin(\omega.t) \Rightarrow \begin{cases} |x(t)| = \sqrt{2} \times |\sin(\omega.t)| \\ \angle x(t) = \frac{\pi}{4} [\pi] \end{cases}$$

$$\omega = 2 \times \pi \times 8.751e3 \text{ rad.s}^{-1}.$$

Sampling time=iteration frequency=200e3 Hz.

5 F table size is m=128.

(The expression "e3" means a multiplication by a factor  $10^3$ ).

In Fig. 6 and the following figures, the following applies:

10  $i_{in} = \text{Re}(x); q_{in} = \text{Im}(x)$  [input - x]  
 $i_{pd} = \text{Re}(z); q_{pd} = \text{Im}(z)$  [predistorter out - f(x)]  
 $i_{fe} = \text{Re}(y); q_{fe} = \text{Im}(y)$  [„PA out“ - g(f(x))]  
 (PA = power amplifier).

15 The adaptive predistorter amplifier shown in Fig. 6 comprises a multiplier 5 to which the input signal x and the output of a look-up table (memory) 6 is applied. The contents of look-up table 6 correspond to the table shown in Fig. 4 or 5. Thus, the output signal of look-up table 6 is the  $F_{i,k}$  value selected  
 20 according to the present amplitude of input signal x. The multiplier 5 generates, as its output, the signal z which is applied to means 2, similar as in Figs. 1 and 3. Here, the means 2 is an amplifier having a time-varying gain characteristic. The output signal y of means 2 and the input  
 25 signal x are supplied to a processing means 7 which stores and processes the above described adaptation algorithm. Updates for the indices i, j, etc..., i.e. F values  $F_{i,k+1}, F_{j,k+1}$ , etc... for the present iteration step "k+1" are calculated within processing means 7 based on the algorithm and are then supplied  
 30 from processing means 7 to the look-up table 6 for storage therein, replacing the previous respective values (see Fig. 4 or 5).

The complex input x is not only supplied to processing means 7

but also to a squaring means 8 which generates the square value of the absolute value of input  $x$ . The output of squaring means 8 is connected to an input of look-up table 6. The present output value of squaring means 8 is used as address (index i) 5 for selecting the  $F_{i,k}$  value to be updated, so that same is selected depending on the square of the absolute value of  $x$ . This selected F value and the at least one further F value (see Fig. 4 or 5) are then updated according to the calculation results of processing means 7.

10 The characteristic of  $g$  (means 2) is shown in Fig. 7. The upper diagram of Fig. 7 represents the magnitude response of  $g$  (horizontal axis: absolute value of input signal  $z$ ; vertical axis: output  $y = g(|z|)$ ) whereas the lower diagram illustrates 15 the phase response of  $g$  (horizontal axis: absolute value of input signal  $z$ ; vertical axis: phase of  $g(|z|)$  in radians). These curves correspond to typical measured data for a given RF power amplifier (data has been normalised). Typically, such an amplifier will produce essentially AM to AM and AM to PM 20 distortion.

For reference, Fig. 8 shows the input and output signals for a non-linearised function  $g$  (i.e.,  $\forall x \in C, f_k(x) = x \Leftrightarrow \forall (i, k) \in [1 \rightarrow m] \times N, F_{i,k}(x) = 1 + j \cdot 0 \Rightarrow y = g(f(x)) = g(x)$ ) with predistortion 25 being switched off. The upper diagram of Fig. 8 illustrates the time-behaviour of the real parts of input  $x$  ( $i_{in} = \text{Re}(x)$ ) and output  $y$  ( $i_{fe} = \text{Re}(y)$ ) with regard to the number of iterations (horizontal axis). The lower diagram of Fig. 8 illustrates the time-behaviour of the imaginary parts of input  $x$  ( $q_{in} = \text{Im}(x)$ ) 30 and output  $y$  ( $q_{fe} = \text{Im}(y)$ ) with regard to the number of iterations (horizontal axis). Here, we are aiming at a gain factor of 1 so the input signals are similar to the ideally expected output signals.

35 Let us first consider a case where the function  $g$  is unknown but time-invariant. Its transfer function is shown in Fig. 7.

for the presented algorithm. The upper diagram shows the imaginary part of input  $x$  over the number of iterations whereas the lower diagram shows the imaginary parts of signals  $z$  and  $y$ .

5 It can be seen from the simulations that the presented algorithm presents faster convergence than the linear method or the LMS method.

10 Next, let us consider a case wherein  $g$  increases monotonically with time.

The new unknown transfer function  $g_{1.2}$  is defined as a ramp of slope 1 multiplied by the previous transfer function  $g$ :  
 $g_{1.2}(t, x) = g(t \times x)$  where  $g$  is the same function as defined in Fig.  
15 7.

Again, Fig. 14 presents the results for this time-varying function  $g$  (ramp) for all three methods, with the inscriptions of the horizontal and vertical axes similar to same of Figs. 9,  
20 10.

Here again, the proposed algorithm reveals to converge faster and hence to be more robust against slow variations of the function  $g$  (gain).

25 In the next considered case, the transfer function  $g$  is a time varying function  $g$  in form of a step. This new transfer function  $g_{1.3}$ , is defined as the previous function  $g$  multiplied by a step function:

30 
$$g_{1.3}(t, x) = g(1 \times x) \text{ for } 0 \leq t \leq 0.01 \quad (t=0.01 \Leftrightarrow \text{iteration 2000})$$
$$g_{1.3}(t, x) = g(1.2 \times x) \text{ for } 0.01 < t$$

35 Fig. 15 presents the results for all three methods.

This case attempts to test the response of the algorithm for sudden variations of  $g$ . The proposed method again converges the quickest. Note that the LMS method is quite slow to readapt.

5 A comparison for  $g$  as a step (same as above) is done for the linear method and the LMS method for the three following algorithms (Figs. 16 and 17):

- 1) Updating current  $x$  address only (index i);
- 10 2) Updating current  $x$  (index i) and  $x_{update}$  (index j) addresses;
- 3) Updating current  $x$  (index i),  $x_{update}$  (index j) and all  $x_{interpolate}$  addresses between  $x$  and  $x_{update}$  (indices w with  $w \in [j \rightarrow i]$  if  $j \leq i$  for example).

15

The algorithm 1) is the known one whereas algorithms 2) and 3) are in accordance with the present invention.

Fig. 16 shows the three updating procedures using the linear 20 method, whereas Fig. 17 shows the three updating procedures using the LMS method. The upper diagram illustrates algorithm 1), the middle diagram algorithm 2), and the lower diagram algorithm 3).

25 It is obvious that algorithm 2) has a better convergence speed than algorithm 1), and that algorithm 3) yet improves this speed dramatically. Interpolation used in algorithm 3) relies on the assumption of a „well-behaved“ response for  $g$ , which becomes even more plausible as the algorithm converges. In 30 general, for a power amplifier, this is a fair assumption.

The algorithm described here thus provides, compared to other 35 algorithms used for predistortion systems such as the linear method or the LMS method, very fast convergence, and is quite robust against fluctuations of  $g$  in time. Besides, it requires few operations (i.e. is computationally simple), and can be

used in conjunction with other algorithms as an enhancement.

The use of spectrally more efficient modulation schemes (higher throughput/occupied bandwidth ratio) to accommodate higher data rates in communications often introduces both phase and magnitude variations of the signal, thus requiring stringent linearity of the transceiver to preserve signal integrity. Transmitter linearisation, in particular, is then often necessary to provide good linearity while offering reasonable power efficiency.

Satellite and mobile applications are two potential candidates for linearisation. The proposed algorithm can, for instance, be used within digital predistortion systems as a linearisation means. The present invention can therefore be embodied in a stationary or mobile device for mobile communications, in particular in a RF transmitter thereof, in such standards as GSM EDGE (Global System for Mobile Communications Enhanced Data Rates for GSM Evolution), TETRA (Terrestrial Trunked Radio), CDMA (Code Division Multiple Access), W-CDMA (Wideband-CDMA), NADC (North American Digital Cellular), PDC (Personal Digital Cellular)...

An embodiment is therefore a stationary or mobile apparatus for mobile communications containing a device as shown in Fig. 3 or 6, preferably as defined in the system claim, and/or operating as defined in the independent method claims.

## Claims

1. Method for compensating deviations of an unknown transfer function from an expected transfer function, and/or for compensating time-varying changes of a transfer function acting on an input signal, for minimising errors of an output signal generated in dependence on the input signal, which input signal is subjected to a first, adaptive transfer function and to a second, unknown and/or varying transfer function to generate the output signal, the first transfer function being updated for compensating deviations or changes of the second transfer function,  
wherein, when updating one point of the first transfer function for a current input signal value, the first transfer function is also updated for at least one other point corresponding to a different input signal value.
2. Method according to claim 1, wherein the at least one other point of the first transfer function is calculated in dependence on the current input signal value and the value of the first transfer function at this input signal value.
3. Method according to claim 1 or 2, wherein the at least one other point of the first transfer function is calculated in additional dependence on the respective value of the second transfer function.
4. Method according to claim 1, 2, or 3, wherein the at least one other point of the first transfer function is corrected in dependence on the value of the second transfer function for the current input signal value.

5. Method according to claim 4, wherein the at least one other point of the first transfer function is corrected by multiplying the value of the second transfer function for the  
5 current input signal value, with a factor.

6. Method according to any one of the preceding claims, wherein values of the first transfer function lying adjacent to the one  
10 point of the first transfer function and/or the at least one other point of the first transfer function are also corrected with appropriate weighting.

15 7. Method according to claim 6, wherein values of the first transfer function lying between the one and the at least one other point of the first transfer function are corrected by interpolating.

20 8. Method according to any one of the preceding claims, wherein the values of the first transfer function are provided in form of a table being addressed by the input signal values.

25 9. Method according to any one of the preceding claims, wherein the updating of the first transfer function is done in an iterative manner.

30 10. System for compensating deviations of an unknown transfer function from an expected transfer function, and/or for compensating time-varying changes of a transfer function acting on an input signal, for minimising errors of an output signal  
35 generated in dependence on the input signal, which input signal is applied via a first, adaptive compensating means having a first, adaptive transfer function, to a second means having a

second, unknown and/or varying transfer function to generate the actual output signal, the first transfer function being updated for compensating deviations and/or changes of the second transfer function by means of a processing means,  
5 wherein the processing means is adapted to update, when updating one point of the first transfer function for a current input signal value, the first transfer function also at least one other point corresponding to a different input signal value.

10

11. Use of the method according to anyone of claims 1 to 9, or of the system according to claim 10, to perform linearisation of a signal path.

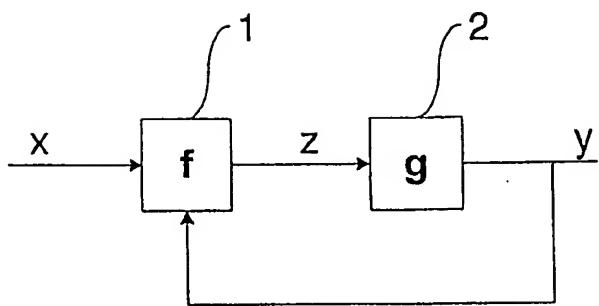


Figure 1: The functional system

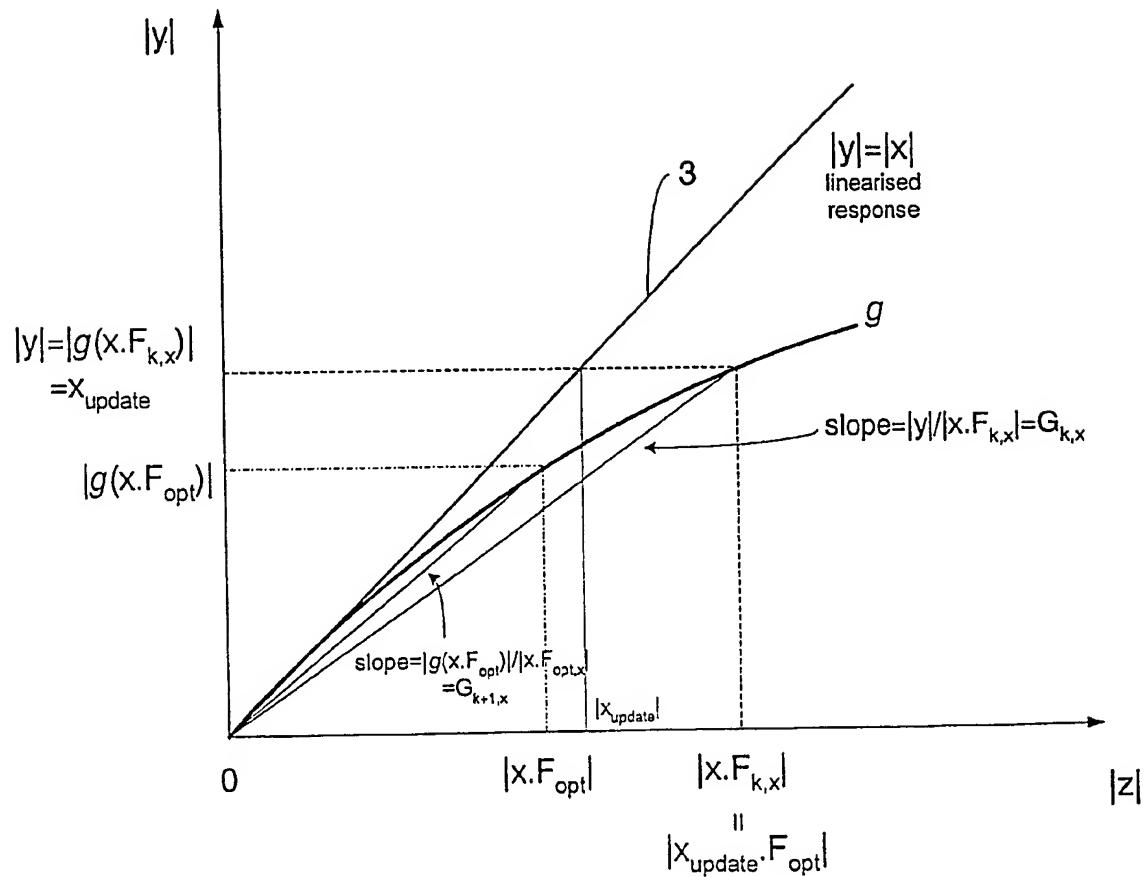


Figure 2: Illustration of the algorithm

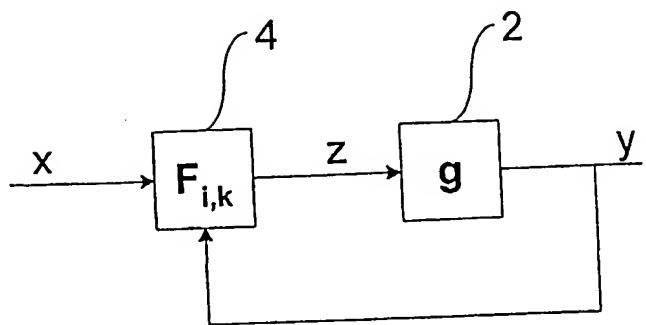


Figure 3: Practical system

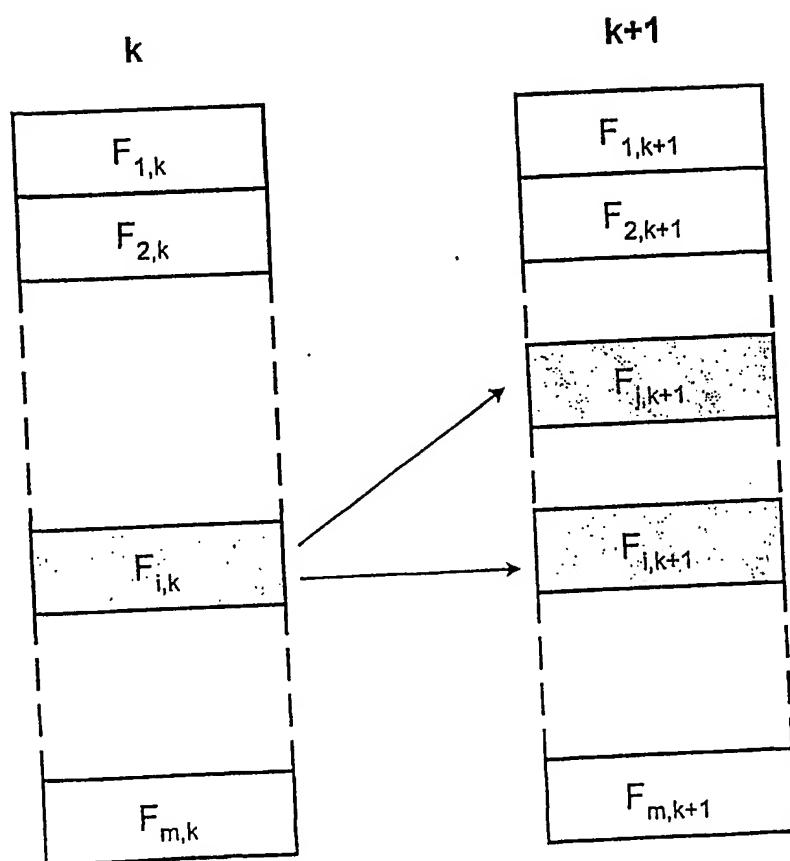


Figure 4: F table

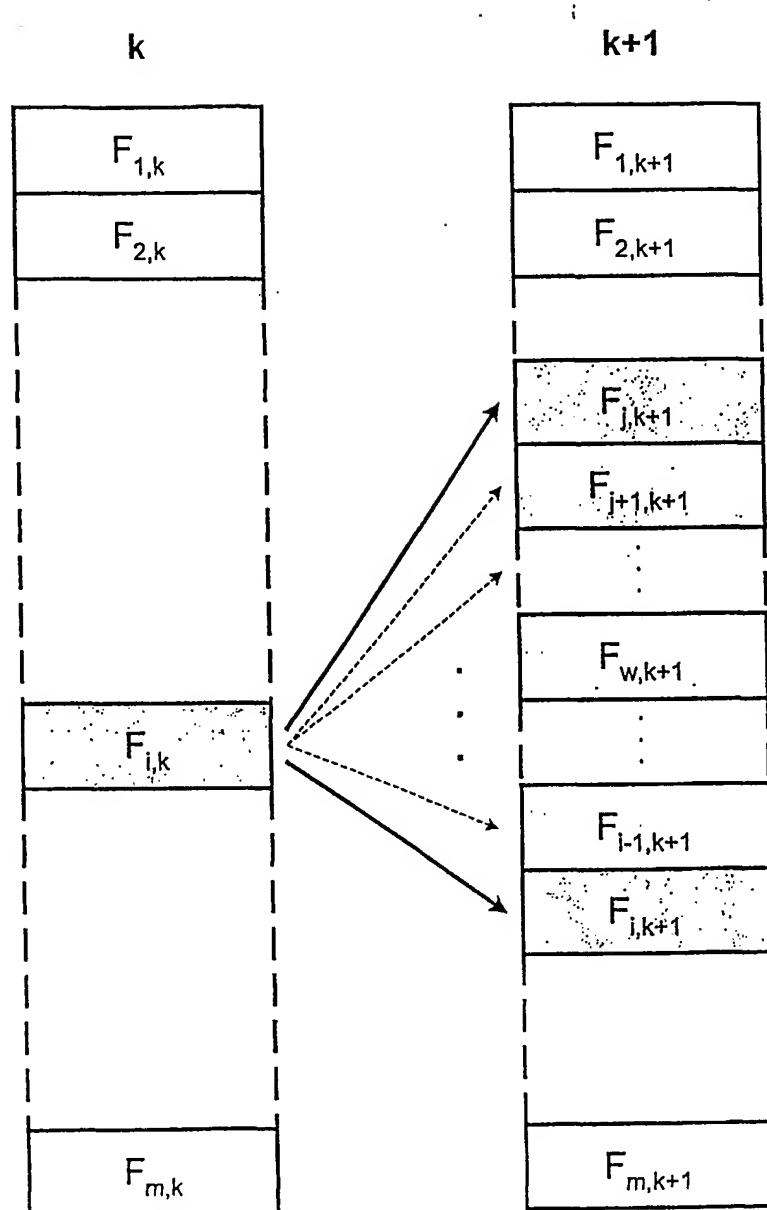


Figure 5: Variation for F table update

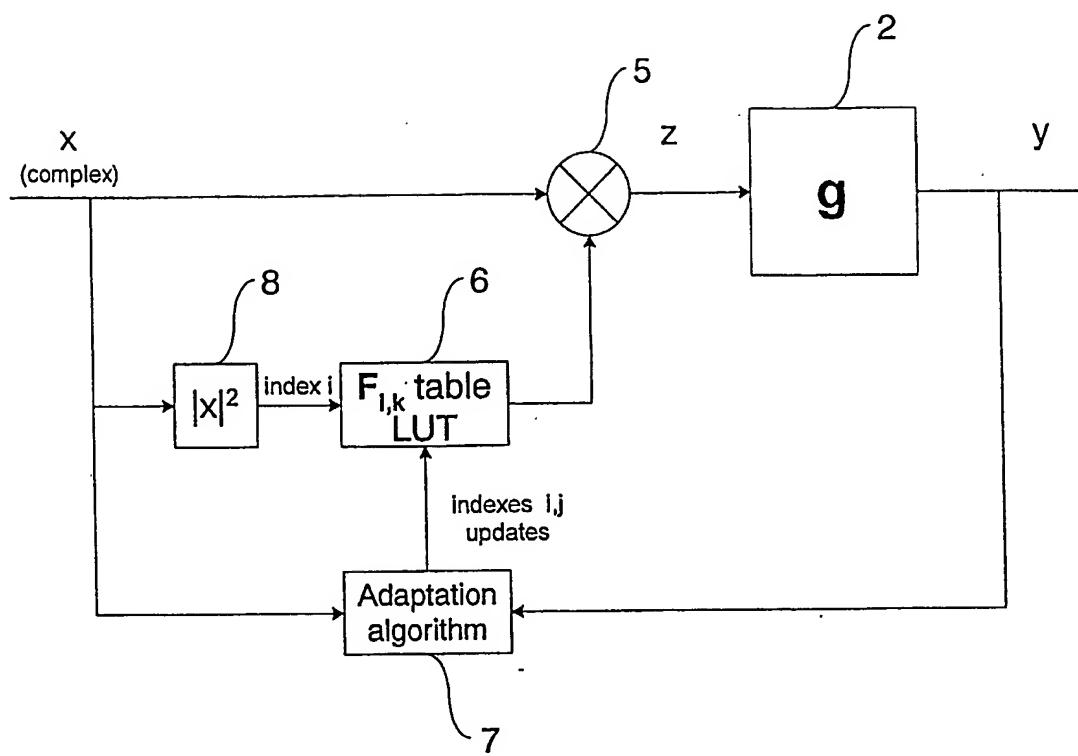


Figure 6: Predistortion block diagram

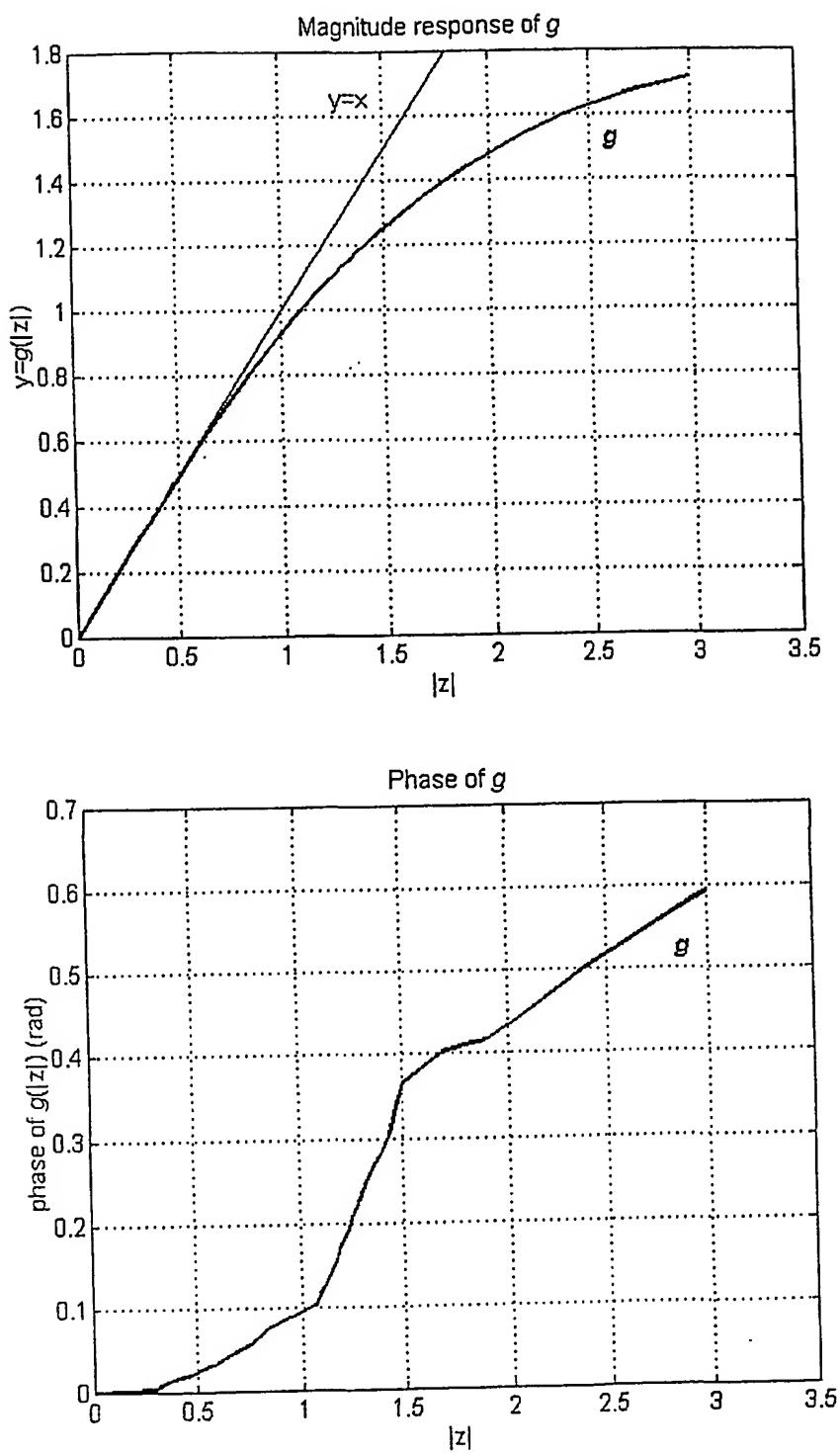


Figure 7: Magnitude and phase response of  $g$

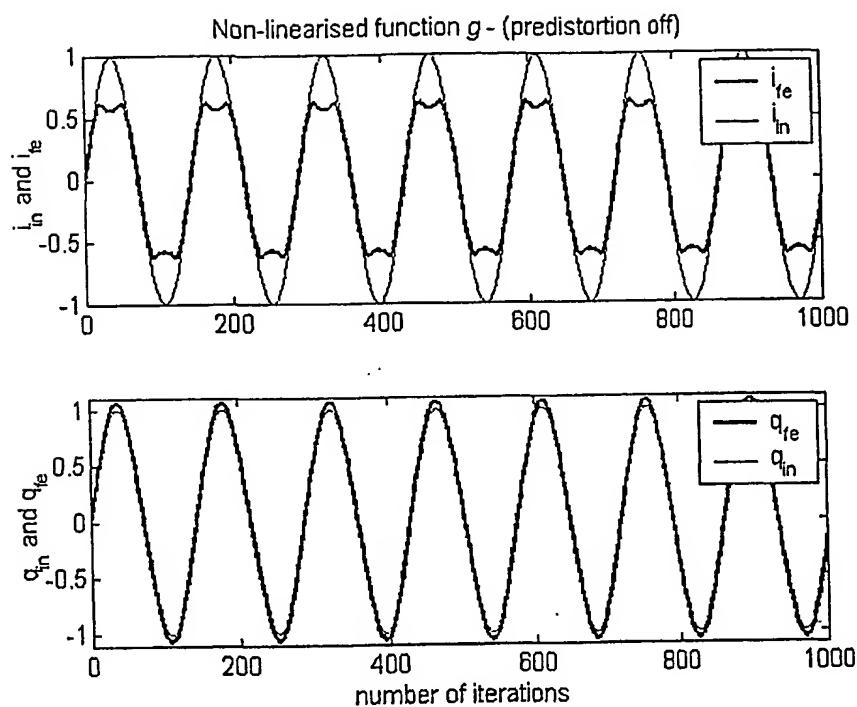


Figure 8: Non-linearised function  $g$  - (predistortion off)

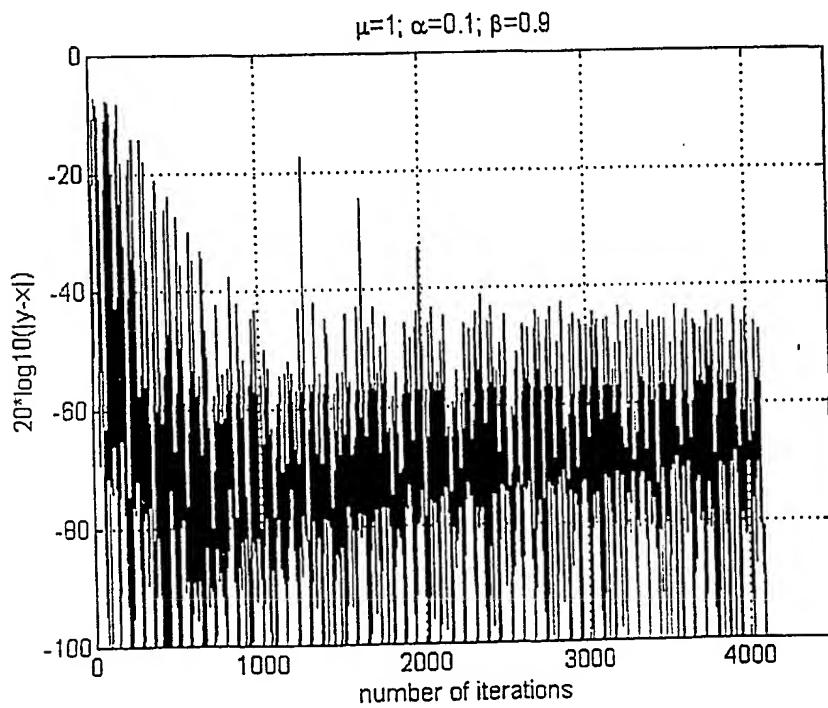


Figure 9:  $20 \log_{10}(|y-x|)$  for the presented algorithm

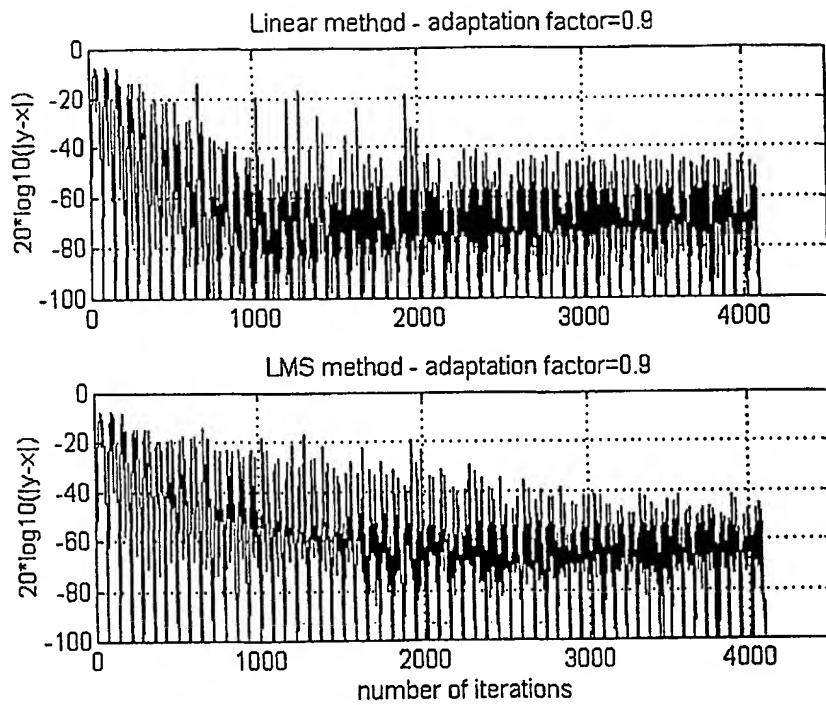


Figure 10:  $20 \cdot \log_{10}(|y-x|)$  for linear and LMS methods

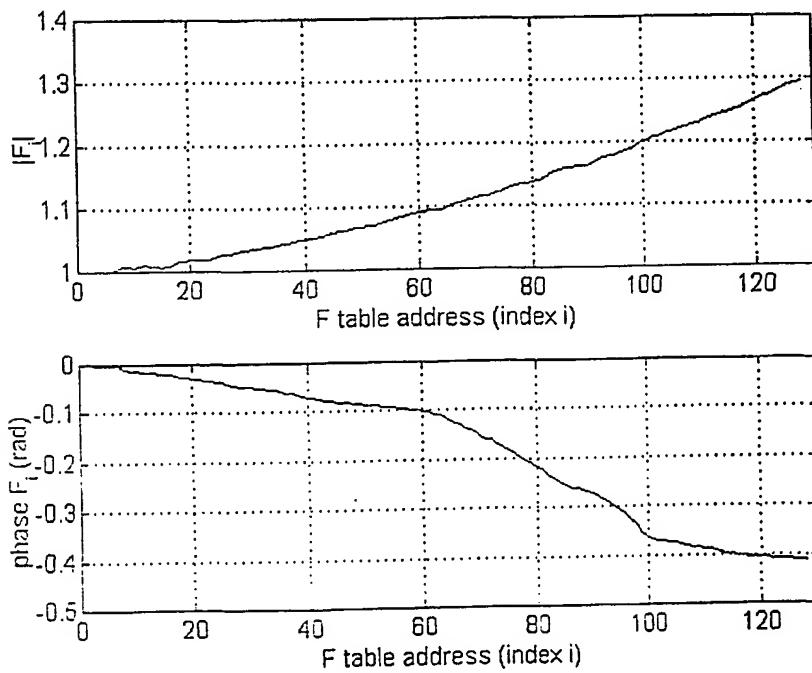


Figure 11:  $F_i$  values at the end of the simulation for the presented algorithm

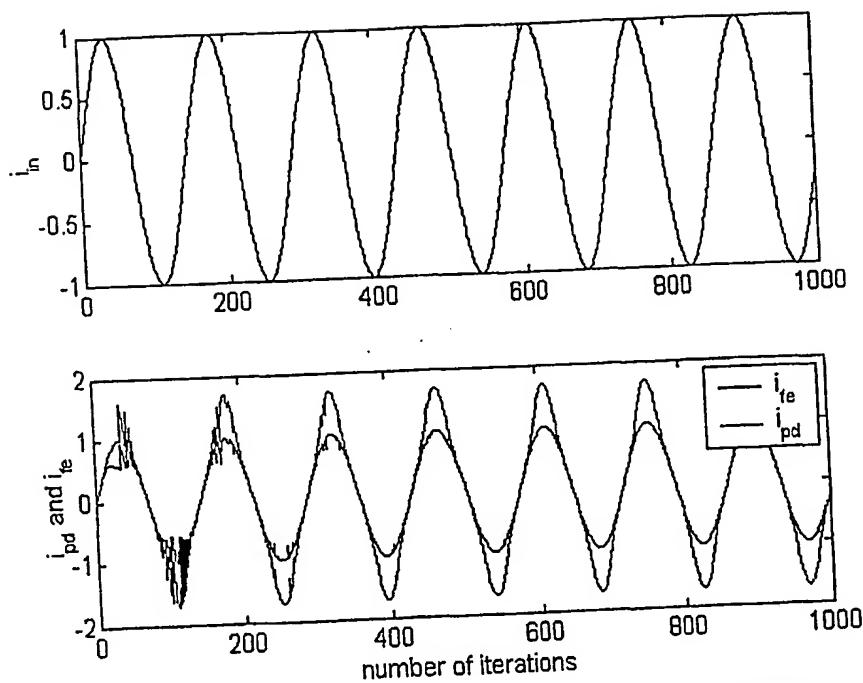


Figure 12: In-Phase components through the chain (input, predistorter out, PA out) for the presented algorithm

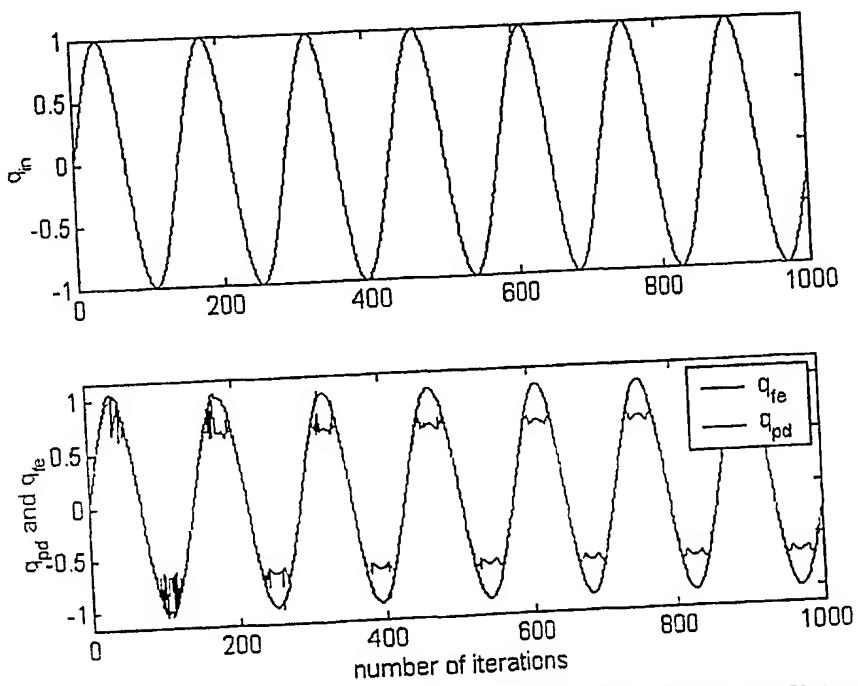


Figure 13: Quadrature components through the chain (input, predistorter out, PA out) for the presented algorithm

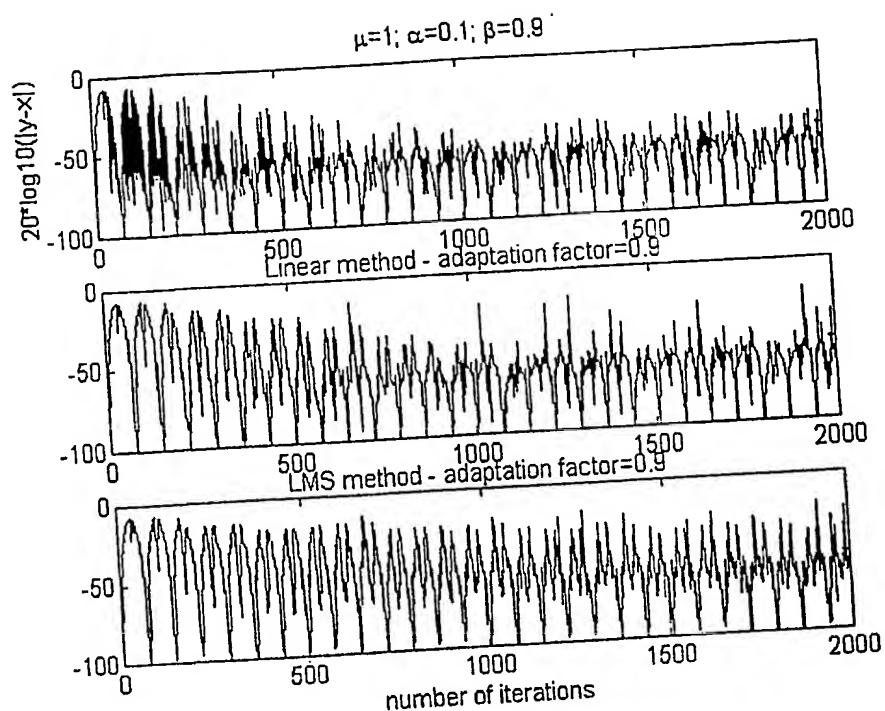


Figure 14: time varying function  $g$  (ramp)

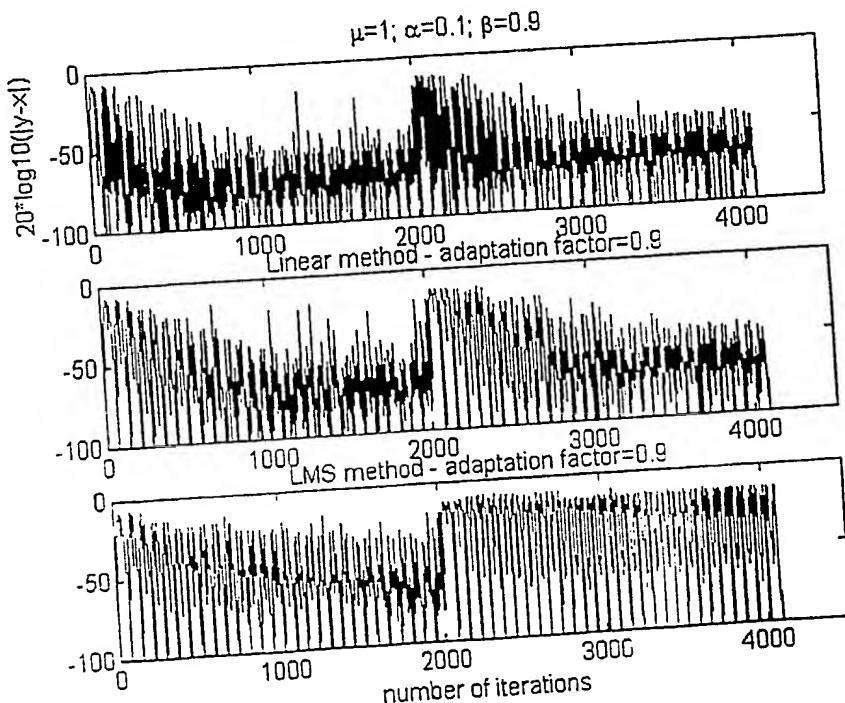


Figure 15: time varying function  $g$  (step)

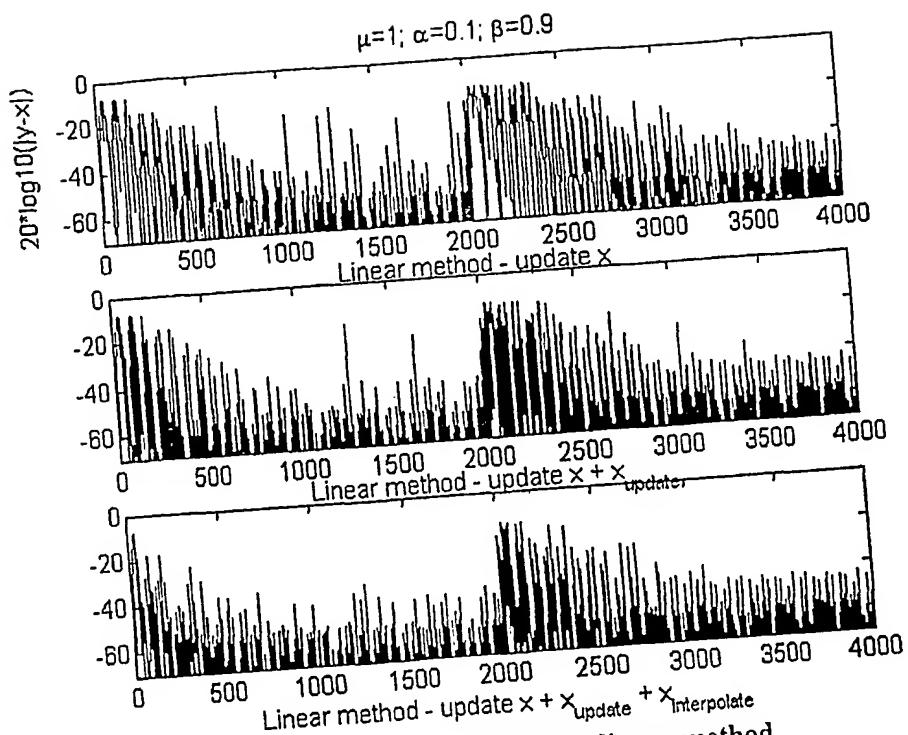


Figure 16: Three updating procedures for the linear method

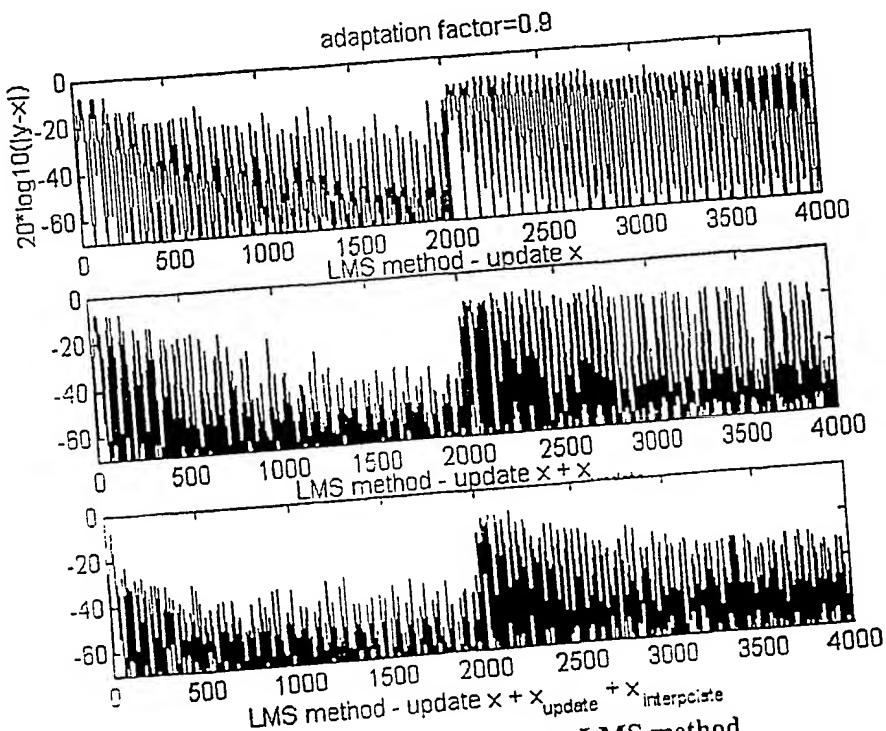


Figure 17: Three updating procedures for the LMS method

## INTERNATIONAL SEARCH REPORT

International Application No  
PCT/EP 00/00579

A. CLASSIFICATION OF SUBJECT MATTER  
IPC 7 H03F1/32

According to International Patent Classification (IPC) or to both national classification and IPC

## B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)  
IPC 7 H03F

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the International search (name of data base and, where practical, search terms used)

EPO-Internal, WPI Data, PAJ

## C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	US 5 900 778 A (MOURA JOSEM F ET AL) 4 May 1999 (1999-05-04) the whole document ---	1-4, 6, 8-11
X	EP 0 658 975 A (ALCATEL ITALIA) 21 June 1995 (1995-06-21) page 7, line 1 - line 8; figure 2 ---	1, 4-7
A	US 5 049 832 A (CAVERS JAMES K) 17 September 1991 (1991-09-17) column 11, line 5 - line 27 -----	

Further documents are listed in the continuation of box C.

Patent family members are listed in annex.

## \* Special categories of cited documents :

- \*A\* document defining the general state of the art which is not considered to be of particular relevance
- \*E\* earlier document but published on or after the international filing date
- \*L\* document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)
- \*O\* document referring to an oral disclosure, use, exhibition or other means
- \*P\* document published prior to the international filing date but later than the priority date claimed

\*T\* later document published after the International filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention.

\*X\* document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone

\*Y\* document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art.

\*&\* document member of the same patent family

Date of the actual completion of the International search

29 September 2000

Date of mailing of the International search report

06/10/2000

Name and mailing address of the ISA

European Patent Office, P.O. Box 5618 Patentlaan 2  
NL - 2280 HV Rijswijk  
Tel. (+31-70) 340-2040, Tx. 31 651 epo nl,  
Fax: (+31-70) 340-3016

Authorized officer

Segaert, P

## INTERNATIONAL SEARCH REPORT

Int'l Application No  
PCT/EP 00/00579

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
US 5900778 A	04-05-1999	NONE	
EP 0658975 A	21-06-1995	IT 1265271 B CA 2137994 A DE 69425317 D JP 8051320 A US 5524286 A	31-10-1996 15-06-1995 24-08-2000 20-02-1996 04-06-1996
US 5049832 A	17-09-1991	NONE	